

Coprime Arrays and Samplers for Space-Time Adaptive Processing

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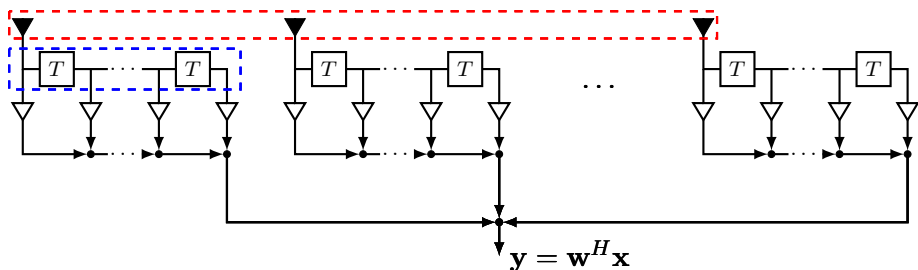
Outline

- 1 Introduction
- 2 Joint Angle-Doppler Estimation using Coprime Arrays and Coprime Samplers (Coprime JADE)
- 3 Numerical Results
- 4 Conclusion

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Space-Time Adaptive Processing¹



- spatial arrays + time sampling.
- STAP resolves the direction-of-arrival (DOA) information and the Doppler frequency jointly.
- Distinguish at most $O(\# \text{ of sensors} \times \# \text{ of taps})$ sources

¹J. R. Guerci, *Space-Time Adaptive Processing for Radar*. Artech House, 2003.

Different Estimators

- Power spectrum density (PSD)²,
- Minimum variance distortionless response (MVDR)³,
- MUSIC⁴,
- ESPRIT⁵,
- compressive joint angular-frequency power spectrum estimation⁶,
- etc...
- Question: Can we estimate more sources than the existing methods?

²R. Klemm, *Space-Time Adaptive Processing: Principles and Applications*. IEEE Press, 1998.

³J. Capon, "High-resolution frequency-wavenumber spectrum analysis," *Proc. IEEE*, vol. 57, no. 8, pp. 1408–1418, 1969.

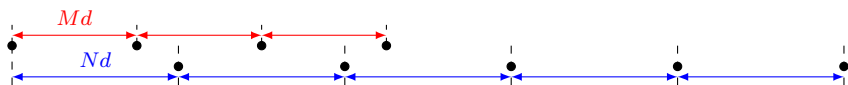
⁴R. Schmidt, "Multiple emitter location and signal parameter estimation," *IEEE Trans. Antennas Propag.*, vol. 34, no. 3, pp. 276–280, 1986.

⁵R. Roy and T. Kailath, "Esprit-estimation of signal parameters via rotational invariance techniques," *IEEE Trans. Acoust., Speech, Signal Process.*, vol. 37, no. 7, pp. 984–995, 1989.

⁶D. D. Ariananda and G. Leus, "Compressive joint angular-frequency power spectrum estimation," in *Proc. the 21st European Signal Processing Conference (EUSIPCO 2013)*, Marrakech, Morocco, 2013.

Coprime Arrays⁷

Physical array:



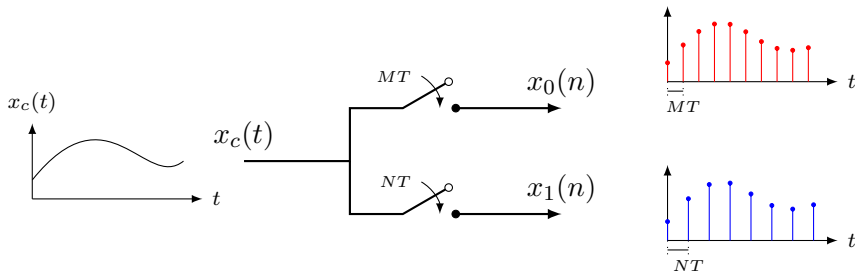
Difference coarray (non-negative part):



- M and N are a coprime pair of integers.
- Union of two uniform linear arrays with interelement spacing Md and Nd , respectively.
- Coprime arrays identify up to $O(MN)$ sources using $O(M + N)$ sensors.

⁷P. P. Vaidyanathan and P. Pal, "Sparse sensing with co-prime samplers and arrays," *IEEE Trans. Signal Process.*, vol. 59, no. 2, pp. 573–586, 2011.

Coprime Samplers in Time⁸



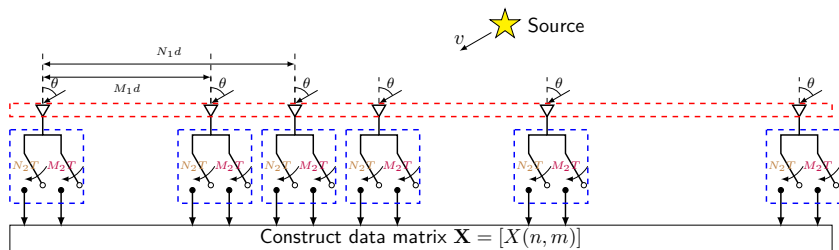
- Non-uniform samples of $x_c(t)$ ($x_0(n)$ and $x_1(n)$) suffice to compute uniform samples $R_c(nT)$, which is the autocorrelation function of $x_c(t)$.
- Reduced sample rate: from $\frac{1}{T}$ to $\frac{1}{MT} + \frac{1}{NT}$

⁸P. P. Vaidyanathan and P. Pal, "Sparse sensing with co-prime samplers and arrays," *IEEE Trans. Signal Process.*, vol. 59, no. 2, pp. 573–586, 2011.

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Coprime JADE



- **Coprime arrays with parameter M_1, N_1** : located at nd , where $n \in \mathbb{S}_s = \{0, M_1, \dots, (N_1 - 1)M_1, N_1, 2N_1, \dots, (2M_1 - 1)N_1\}$.
- **Coprime samplers with parameter M_2, N_2** : samples at $t = mT$, where $m \in \mathbb{S}_t = \{0, M_2, \dots, (N_2 - 1)M_2, N_2, 2N_2, \dots, (2M_2 - 1)N_2\}$.

The Data Model (Matrix forms)

The data matrix \mathbf{X} can be modeled as

$$\mathbf{X} = \sum_{i=1}^D A_i \mathbf{v}_s(\bar{\theta}_i) \mathbf{v}_t^T(\bar{f}_i) + \mathbf{N}, \quad (1)$$

where

- A_i : The complex amplitude of the i th source.
- $\mathbf{v}_s(\bar{\theta}_i) = \left[e^{j2\pi\bar{\theta}_i n} \right]_{n \in \mathbb{S}_s}$: Spatial steering vectors.
- $\mathbf{v}_t(\bar{f}_i) = \left[e^{j2\pi\bar{f}_i m} \right]_{m \in \mathbb{S}_t}$: Temporal steering vectors.
- $\bar{\theta}_i = \frac{d}{\lambda} \sin \theta_i$: Normalized DOA. θ_i : DOA.
- $\bar{f}_i = \frac{T}{\lambda} v_i$: Normalized Doppler frequency. v_i : radial velocity.
- \mathbf{N} : Additive noise.

The Data Model (Vector forms)

Vectorizing \mathbf{X} gives

$$\mathbf{x} = \text{vec}(\mathbf{X}) = \sum_{i=1}^D A_i \mathbf{v}_{s,t}(\bar{\theta}_i, \bar{f}_i) + \mathbf{n}. \quad (2)$$

- $\mathbf{v}_{s,t}(\bar{\theta}_i, \bar{f}_i) = \mathbf{v}_t(\bar{f}_i) \otimes \mathbf{v}_s(\bar{\theta}_i)$: Space-time steering vectors.
- \otimes : Kronecker products.
- Convert (2) to the coarray-lag domain.

Representation in the Coarray-Lag Domain

- Statistical assumptions:

- $A_1, \dots, A_D, \mathbf{n}$ are **zero mean, uncorrelated** random variables/vectors.
- For $i = 1, \dots, D$, $\mathbb{E} [|A_i|^2] = \sigma_i^2$, $\mathbb{E} [\mathbf{nn}^H] = \sigma^2 \mathbf{I}$.

- The covariance matrix of \mathbf{x} :

$$\mathbf{R}_{\mathbf{x}} = \mathbb{E} [\mathbf{xx}^H] = \sum_{i=1}^D \sigma_i^2 \underbrace{\mathbf{v}_{s,t}(\bar{\theta}_i, \bar{f}_i) \mathbf{v}_{s,t}^H(\bar{\theta}_i, \bar{f}_i)}_{(\mathbf{v}_t(\bar{f}_i) \mathbf{v}_t^H(\bar{f}_i)) \otimes (\mathbf{v}_s(\bar{\theta}_i) \mathbf{v}_s^H(\bar{\theta}_i))} + \sigma^2 \mathbf{I}.$$

- $\mathbf{R}_{\mathbf{x}}$ can be reshaped into another matrix \mathbf{Z} :

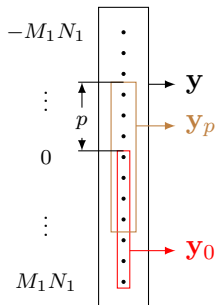
$$\mathbf{Z} = \sum_{i=1}^D \sigma_i^2 \mathbf{w}_s(\bar{\theta}_i) \mathbf{w}_t^T(\bar{f}_i) + \sigma^2 \mathbf{e}_1 \mathbf{e}_2^T. \quad (3)$$

- $\mathbf{w}_s(\bar{\theta}_i)$: Spatial steering vector over **the different set of \mathbb{S}_s (Coarrays)**.
- $\mathbf{w}_t(\bar{f}_i)$: Temporal steering vector over **the difference set of \mathbb{S}_t (Lags)**.

Spatial Smoothing

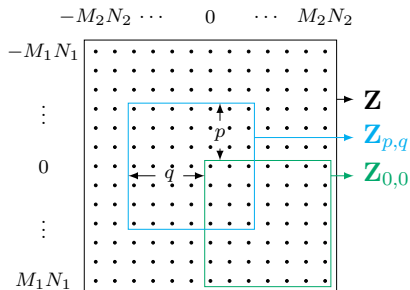
- Establish a full-rank matrix \mathbf{R}_{ss} from \mathbf{Z} and then apply the MUSIC algorithm.
- How to perform spatial smoothing over the **matrix \mathbf{Z}** ?

1D Spatial Smoothing



$$\mathbf{R}_{ss} \propto \sum_p \mathbf{y}_p \mathbf{y}_p^H$$

2D Spatial Smoothing



Remarks

- Coprime JADE is similar to 2D coprime arrays.
 - (DOA, Doppler) \equiv (Azimuth, Elevation).
 - Coprime JADE: Fixed sensor locations and the same sampling pattern is applied to each sensor.
 - 2D coprime arrays: Design over a nonseparable lattice.
- Let $\mathbf{R}_{ss} = [\mathbf{U}_s \ \mathbf{U}_n] \mathbf{\Lambda} [\mathbf{U}_s \ \mathbf{U}_n]^H$ as the eigen-decomposition, where \mathbf{U}_n represents the *noise subspace*. The MUSIC spectrum is

$$P_{MUSIC}(\bar{\theta}, \bar{f}) = \frac{1}{\|\mathbf{U}_n^H \tilde{\mathbf{w}}_{s,t}(\bar{\theta}, \bar{f})\|_2^2},$$

where $\bar{\theta}, \bar{f} \in [-1/2, 1/2)$ and $\|\cdot\|_2$ denotes Euclidean norms of vectors.

The Number of Identifiable Sources⁹

Theorem

Consider distinct sources $S = \{(\bar{\theta}_i, \bar{f}_i)\}_{i=1}^D$ where $D \leq M_1 N_1 M_2 N_2$.

Assume that

- the set $\{\bar{\theta}_i\}_{i=1}^D$ takes at most $M_1 N_1$ distinct values and
- $\{\bar{f}_i\}_{i=1}^D$ contains at most $M_2 N_2$ distinct values.

Then, $P_{MUSIC}(\bar{\theta}, \bar{f})$ has a singularity if and only if $(\bar{\theta}, \bar{f}) \in S$.

Intuition:

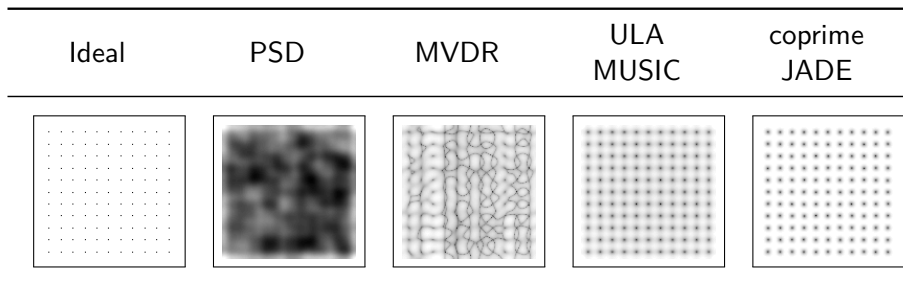
- Coprime arrays identify up to $M_1 N_1$ DOAs.
- Coprime samplers identify up to $M_2 N_2$ frequencies.

⁹P. Pal and P. P. Vaidyanathan, "Nested arrays: a novel approach to array processing with enhanced degrees of freedom," *IEEE Trans. Signal Process.*, vol. 58, no. 8, pp. 4167–4181, 2010.

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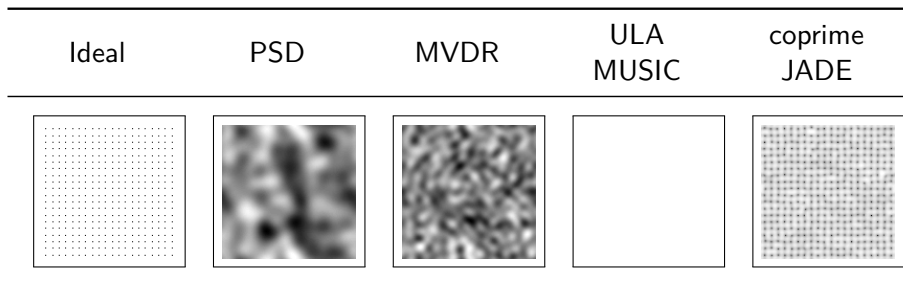
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Angle-Doppler patterns: 121 Sources



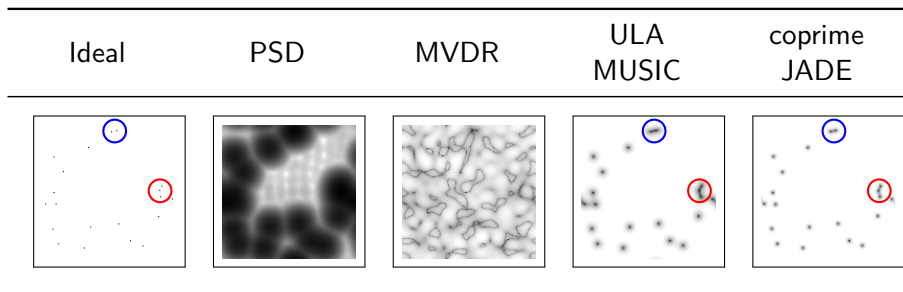
- 11×11 uniformly-placed sources, 1000 snapshots and 0dB SNR.
- $M_1 = M_2 = 4$ and $N_1 = N_2 = 5$ so that the number of sensors and the number of samples per sensor are 12 in all methods.
- Performance limit of ULA + uniform sampling.

Angle-Doppler patterns: 400 Sources



- 20×20 uniformly-placed sources, 1000 snapshots and 0dB SNR.
- $M_1 = M_2 = 4$ and $N_1 = N_2 = 5$ so that the number of sensors and the number of samples per sensor are 12 in all methods.
- Performance limit of coprime array + coprime sampling.

Angle-Doppler patterns: 20 Sources



- 20 randomly-placed sources, 1000 snapshots and 0dB SNR.
- $M_1 = M_2 = 4$ and $N_1 = N_2 = 5$ so that the number of sensors and the number of samples per sensor are 12 in all methods.
- Coprime JADE has the best performance in terms of spread-out.

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Conclusion

- Coprime JADE = Coprime arrays + Coprime samplers.
- Coprime JADE identifies up to $O(M_1 N_1 M_2 N_2)$ sources.
 - STAP only achieves $O((M_1 + N_1)(M_2 + N_2))$ degrees of freedom.
- Future work:
 - Connection between coprime JADE and coprime arrays in high dimensions¹⁰.
 - Different coprime sampling schemes at different sensors.

Thank you for your attention!

¹⁰P. P. Vaidyanathan and P. Pal, "Theory of sparse coprime sensing in multiple dimensions," *IEEE Trans. Signal Process.*, vol. 59, no. 8, pp. 3592–3608, 2011.